

Discrete Fourier Transform

- **Roots of Unity**

$$\omega = e^{\frac{2\pi}{N}nj}, \quad n = 0, \dots, N-1$$

$$\omega_N = e^{\frac{2\pi}{N}j}$$

$$\boxed{\sum_{n=0}^{N-1} \omega_N^n = 0}$$

- **Conjugate pair**

$$\omega^N = 1 \implies (\bar{\omega})^N = 1$$

- **DFT Basis**

Define basis $\{u_0, \dots, u_{N-1}\}$ for \mathbb{C}^N as:

$$u_k[n] = \frac{1}{\sqrt{N}} \omega_N^{kn}$$

Theorem 1 (DFT basis is orthonormal).

Proof.

$$\begin{aligned} \langle u_k, u_{k'} \rangle &= \sum_{n=0}^{N-1} u_k[n] \overline{u_{k'}[n]} \\ &= \frac{1}{N} \sum_{n=0}^{N-1} \omega_N^{kn} \omega_N^{-k'n} \\ &= \frac{1}{N} \sum_{n=0}^{N-1} \omega_N^{(k-k')n} \\ &= \frac{1}{N} \sum_{n=0}^{N-1} \zeta^n \end{aligned}$$

□

Definition 1 (Synthesis). *The change of coordinates from the DFT basis, $F^* \in \mathbb{C}^{n \times n}$.*

$$F_{kn}^* = \frac{1}{\sqrt{N}} \omega_N^{kn}$$

$$F^* = \frac{1}{\sqrt{N}} \begin{pmatrix} 1 & 1 & \dots & 1 \\ 1 & \omega & \dots & \omega^{N-1} \\ 1 & \vdots & \ddots & \vdots \\ 1 & \omega^{N-1} & \dots & \omega^{(N-1)(N-1)} \end{pmatrix} \quad \text{(Synthesis Matrix)}$$

$$\boxed{x = F^* X} \quad \text{(Synthesis Equation)}$$

Definition 2 (Analysis). *The change of coordinates to the DFT basis, $F \in \mathbb{C}^{n \times n}$.*

$$F_{kn} = \overline{F_{nk}^*} = \frac{1}{\sqrt{N}} \omega_N^{-kn}$$

$$F = \frac{1}{\sqrt{N}} \begin{pmatrix} 1 & 1 & \dots & 1 \\ 1 & \omega^{-1} & \dots & \omega^{-(N-1)} \\ 1 & \vdots & \ddots & \vdots \\ 1 & \omega^{-(N-1)} & \dots & \omega^{-(N-1)(N-1)} \end{pmatrix} \quad \text{(Analysis Matrix)}$$

$$\boxed{X = Fx} \quad \text{(Analysis Equation)}$$

$$X[k] = \sum_{n=0}^{N-1} x[n] e^{-j2\pi \frac{kn}{N}} = \sum_{n=0}^{N-1} x[n] \omega_N^{-kn} \quad \text{analysis}$$

$$x[n] = \frac{1}{N} \sum_{k=0}^{N-1} X[k] e^{j2\pi \frac{kn}{N}} = \frac{1}{N} \sum_{k=0}^{N-1} X[k] \omega_N^{kn} \quad \text{synthesis}$$

$$\boxed{x[n] = \sum_{k=0}^{N-1} X[k] u_k[n]}$$

- **DFT of sinusoid**

$$\begin{aligned}
 x[n] &= \alpha \cos\left(\frac{2\pi k}{N}n + \varphi\right), \quad n = 0, \dots, N-1 \\
 &= \frac{\alpha}{2} \left(e^{j\left(\frac{2\pi k}{N}n + \varphi\right)} + e^{-j\left(\frac{2\pi k}{N}n + \varphi\right)} \right) \\
 &= \frac{\alpha}{2} \left(e^{\frac{2\pi k}{N}nj} e^{j\varphi} + e^{-\frac{2\pi k}{N}nj} e^{-j\varphi} \right) \\
 &= \frac{\alpha}{2} \left(\left(e^{\frac{2\pi j}{N}} \right)^{kn} e^{j\varphi} + \left(e^{\frac{2\pi j}{N}} \right)^{-kn} e^{-j\varphi} \right) \\
 &= \frac{\alpha}{2} \left(\omega_N^{kn} e^{j\varphi} + \omega_N^{-kn} e^{-j\varphi} \right) \\
 &= \frac{\alpha}{2} e^{j\varphi} \omega_N^{kn} + \frac{\alpha}{2} e^{-j\varphi} \omega_N^{-kn} \\
 &= \frac{\alpha\sqrt{N}}{2} e^{j\varphi} u_k[n] + \frac{\alpha\sqrt{N}}{2} e^{-j\varphi} u_{N-k}[n]
 \end{aligned}$$

$$x = \frac{\alpha\sqrt{N}}{2} e^{j\varphi} u_k + \frac{\alpha\sqrt{N}}{2} e^{-j\varphi} u_{N-k}$$

If $k = 0$ modulo N , then $u_k = u_0 = u_{N-k}$:

$$x = \left(\alpha\sqrt{N} \cos \varphi \right) u_0$$

else:

$$X[n] = \begin{cases} \frac{\alpha\sqrt{N}}{2} e^{j\varphi}, & n = k \\ \frac{\alpha\sqrt{N}}{2} e^{-j\varphi}, & n = N - k \\ 0, & \text{else} \end{cases}$$

- **Properties of DFT:**

- Linear:

$$F(ax + by) = aF_x + bF_y = aX + bY$$

- Energy-preserving (**Parseval's Theorem**):

$$\|Fx\|^2 = \|x\|^2$$

- Conjugate-symmetric for real signals: if x is **real**, then

$$X[n] = \overline{X[-n]} = \overline{X[N-n]}$$

DFT of A Square Wave

- Let $x \in \mathbb{C}^N$ be the following rectangular pulse, which approximates a square wave when $M = N/4$

$$x[n] = \begin{cases} 1, & -M \leq n \leq M \\ 0, & \text{else} \end{cases}$$

$$X[n] \approx \frac{\sqrt{N}}{2} \operatorname{sinc}\left(\frac{1}{2}n\right), \quad n = -\left\lfloor \frac{N}{2} \right\rfloor \dots 0 \dots \left\lfloor \frac{N}{2} \right\rfloor$$

- “The DFT of a square is a sinc and the DFT of a sinc is a square.”
- If square wave x and sinc X :

$$X = Fx$$

If both x and X are real,

$$\bar{X} = \overline{Fx} \implies X = \bar{F}x = F^*x$$

$$\boxed{FX = x}$$

Only true when x and X are both **real** and $F \neq F^*$!

Linear Time Invariant (LTI) Systems

- Properties:**

- Linear:

1. Scaling:

If $u[n] \rightarrow y[n]$ and $u_2[n]$, then

$$au[n] \rightarrow ay[n]$$

2. Superposition:

If $u_1[n] \rightarrow y_1[n]$ and $u_2[n] \rightarrow y_2[n]$, then

$$u_1[n] + u_2[n] \rightarrow y_1[n] + y_2[n]$$

- Time Invariant:

$$u[n - n_0] \rightarrow y[n - n_0]$$

- Impulse response

$$\delta[n] \triangleq \begin{cases} 1 & \text{if } n = 0 \\ 0 & \text{otherwise} \end{cases}$$

$$\boxed{y[n] = \sum_m u[m]h[n - m]}$$